In DPLL, frequency division number n will affect  $\phi_e$  and  $\Delta f_s$ , so n must be selected properly to let  $\phi_e$  satisfies the system's requirement and let  $t_s \downarrow$ ,  $t_c \uparrow$ ,  $\Delta f_s \uparrow$ 

#### **Frame Synchronization**

As mentioned, carrier synchronization and symbol synchronization needs to estimate the phase of synchronous signal which can be realized by using a PLL. Frame synchronization is realized in a different way——inserting frame alignment signal (distinctive bit sequence). Therefore, the basic task of frame synchronization is how to detect the alignment symbol.

Besides add frame alignment bits, some code such as selfsynchronizing code can be synchronized without adding extra bits. In this section, we only focus on the first method ——inserting frame alignment signal.

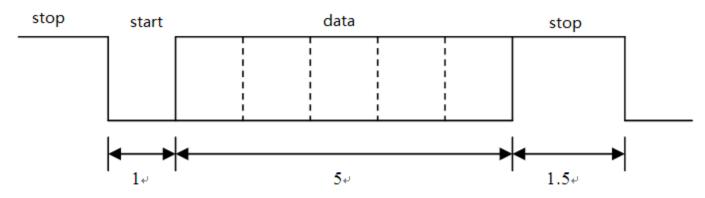
#### **Frame Synchronization**

- Bunched frame alignment signal
- Distributed frame alignment signal

Start-stop Method

#### 1. Start-stop method

It is widely used in teleprinter. Each symbol contains 5-8 data bits, a start bit and a stop bit.



### Start-stop Method

#### Drawbacks:

- 1). Low transport efficiency
- 2). Low precision of timing

# Bunched frame alignment signal

#### 2. Bunched frame alignment signal

This method inserts synchronous code at a particular place in each frame. The code should have a sharp self-correlation function. The detector should be simple to implement.

Frame synchronization code: Barker code, optimal synchronous code and pseudo-random code.

(1) Barker code:

A n bits barker code  $\{x_1, x_2, x_3 \cdots = +1 \text{ or } -1 \}$  its self-correlation function satisfies:

$$R_{x}(j) = \sum_{i=1}^{n-j} x_{i} x_{i+j} = \begin{cases} n & j = 0\\ 0 \text{ or } \pm 1 & 0 < j < n\\ 0 & j \ge n \end{cases}$$

Barker code is not a periodic sequence. It is proved that when n < 12100, we can only find barker code with n = 2, 3, 4, 5, 7, 11, 13.

n	barker code
2	++
3	++-
4	+++-, ++-+
5	+ + + - +
7	++++-
11	++++-
13	++++-++++++++++++++++++++++++++++++++++

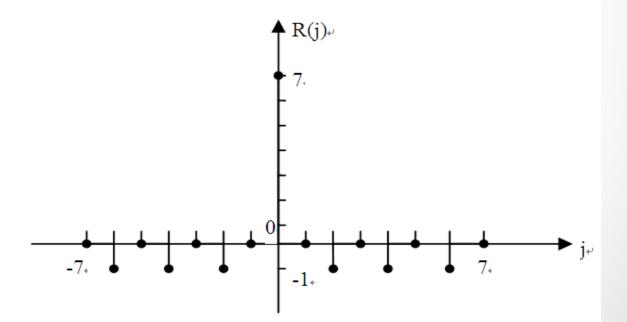
Example: A barker code with n = 7, find its self-correlation function

$$j = 0: R_x(0) = x_1x_1 + x_2x_2 + \cdots = 7$$

$$j = 1: R_x(1) = x_1 x_2 + x_2 x_3 + \cdots$$

Similarly, we can determine  $R_x(j)$ .

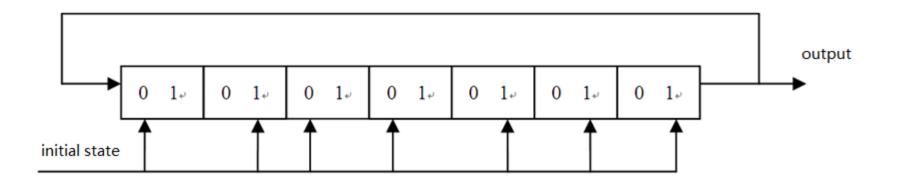
The result is shown below, we can see it has a sharp peak when j = 0.



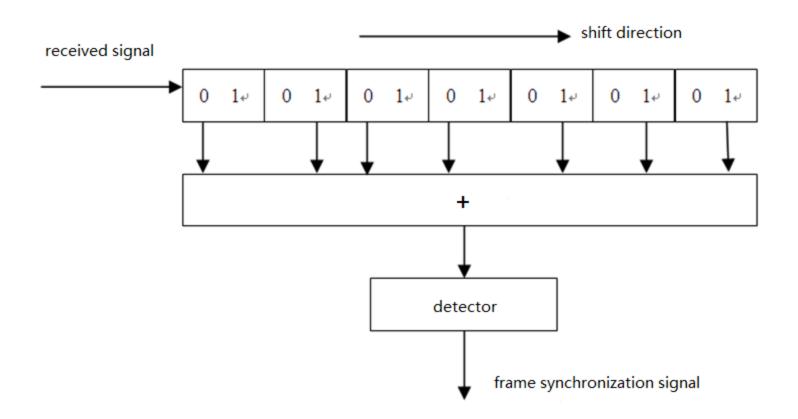
(2) Barker code generator

—shift register

Example: when n=7, a 7 bits shift register. The initial state is a barker code.

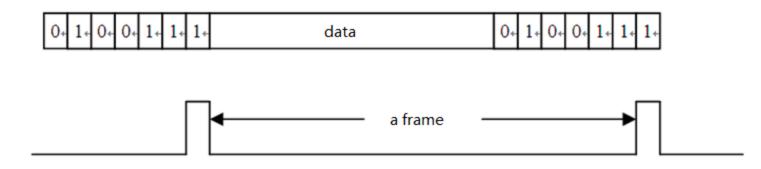


(3) Barker code detector



The barker code detector follows: input:"1" {output "1":? +1 output "0":? -1 input:"0" {output "1":? -1 output "0":? +1

If the output connection of the shift register is the same with a barker code, then when the input is a barker code, the output of the shift register is "1111111". The detector will send a synchronous impulse.



# Distributed frame alignment signal

The synchronous code is distributed in the data signal. That means between each n bits, a synchronous bit is inserted.

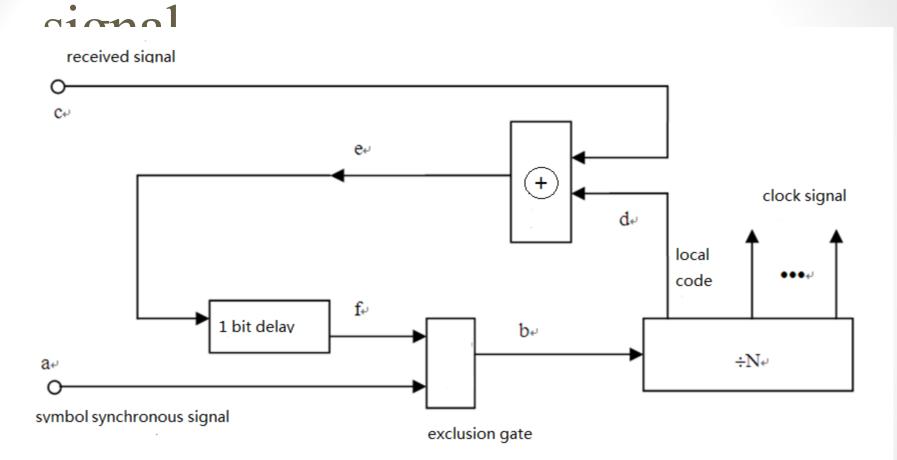
How to design synchronous code:

1. Easy to detect. For example: "11111111" or "10101010"

2. Easy to separate synchronous code and data code. For example: In some digital telephone system, all "0" stands for ring, so synchronous code can only use "10101010".

To determine the synchronous code, receiver need to detect the code bit by bit. Generally, the code can be detected by shifting the signal code by code.

# Distributed frame alignment



# Distributed frame alignment signal

Example: data code is all "0", synchronous code is all "1"

The synchronous code is generated by frequency division (N=4)of the symbol synchronous impulse (a). In practical system, the local code (d) will not be exact the same with (c). Therefore, the output of the XOR gate will have nonzero waveform (e). After one bit delay (f), the exclusion gate will discard one symbol synchronous impulse (b). By repeating this procedure, finally (d) and (c) will be exactly the same, frame synchronization is realized.

Performance of frame synchronization system —Bunched frame alignment signal

#### 1. Probability of missing synchronization $P_L$

Affected by noise, the detector may not be able to detect the synchronous code. The probability of this situation is called probability of missing synchronization  $P_L$ .

Assume the length of synchronous code is n, bit error rate is Pe. The detector will not be able to detect if more than m bit errors happen, then:

$$P_{L} = 1 - \sum_{x=0}^{m} C_{n}^{x} P_{e}^{x} \left(1 - P_{e}\right)^{n-x}$$

#### Performance 2. Probability of false synchronization P<sub>F</sub>

Since data code can be arbitrary, it may be the same with synchronous code. The probability of this situation is called probability of false synchronization  $P_{F}$ .

 $P_{\rm F}$  equals to the probability of appearance of synchronous code in the data code.

a. In a binary code, assume 0 and 1 appears with the same probability. There are  $2^n$  combinations of a n bit code.

b. Assume when there are more than m bit errors, the data code will also be detected as synchronous code.

When m = 0, only  $1(C_n^0)$  code will be detected as synchronous code;

When m = 1, there are  $C_n^1$  codes will be detected as synchronous code;

Therefore, the probability of false synchronization is:

$$P_F = \frac{\sum_{x=0}^{m} C_n^x}{2^n} = \left(\frac{1}{2}\right)^n \sum_{x=0}^{m} C_n^x$$

. . . . . .

 $P_L$  and  $P_F$  depends on the length of synchronous code n and the maximum bit error m. When  $n \uparrow$ ,  $P_F \downarrow$ ,  $P_L \uparrow$ ; when  $m \uparrow$ ,  $P_L \downarrow$ ,  $P_F \uparrow$ 

3. Average build time  $t_s$ 

Assume both  $P_L$  and  $P_F$  will not happen, the worst case is we need one frame to build frame synchronization. Assume each frame contains N bits, each bit has a width  $T_b$ , then one frame costs  $NT_b$ .

Now assume a missing synchronization or a flase synchronization also needs  $NT_b$  to rebuild the synchronization, then:

 $t_{s}^{1} = NT_{b}\left(1 + P_{L} + P_{F}\right)$ 

Bisedes, the average build time of using the distributed frame alignment signal is:

 $t_{s}^{2} = N^{2}T_{b}\left(N \gg 1\right)$ 

Apparently,  $t_s^1 < t_s^2$ , so the previous method is more widely used.