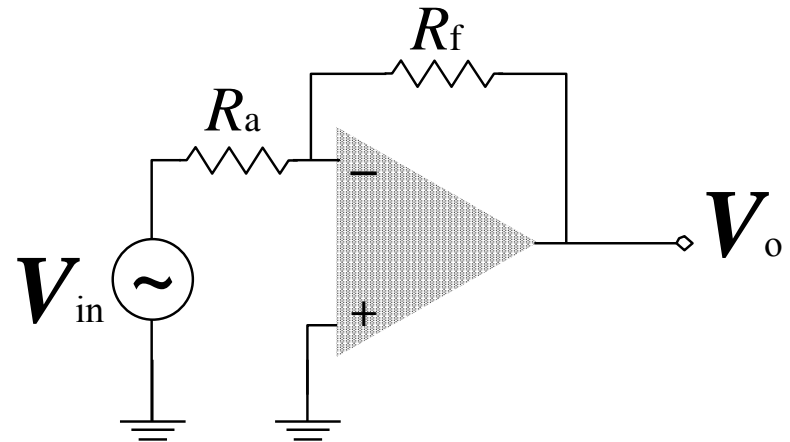


Inverting Amplifier

(1) Kirchhoff node equation at V_+ yields, $V_+ = 0$

(2) Kirchhoff node equation at V_- yields,
$$\frac{V_{in} - V_-}{R_a} + \frac{V_o - V_-}{R_f} = 0$$



(3) Setting $V_+ = V_-$ yields

$$\frac{V_o}{V_{in}} = \frac{-R_f}{R_a}$$

Notice: The **closed-loop gain** V_o/V_{in} is dependent upon the ratio of two resistors, and is independent of the open-loop gain. This is caused by the use of feedback output voltage to subtract from the input voltage.

Multiple Inputs

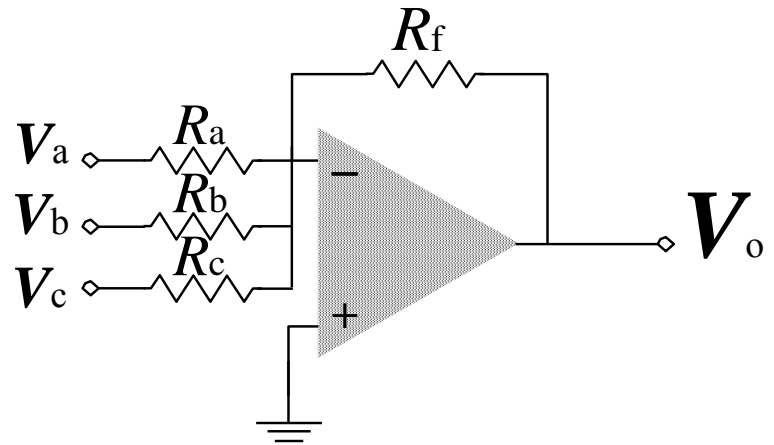
(1) Kirchhoff node equation at V_+ yields, $V_+ = 0$

(2) Kirchhoff node equation at V_- yields,

$$\frac{V_- - V_o}{R_f} + \frac{V_- - V_a}{R_a} + \frac{V_- - V_b}{R_b} + \frac{V_- - V_c}{R_c} = 0$$

(3) Setting $V_+ = V_-$ yields

$$V_o = -R_f \left(\frac{V_a}{R_a} + \frac{V_b}{R_b} + \frac{V_c}{R_c} \right) = -R_f \sum_{j=a}^c \frac{V_j}{R_j}$$



Inverting Integrator

Now replace resistors R_a and R_f by complex components Z_a and Z_f , respectively, therefore

$$V_o = \frac{-Z_f}{Z_a} V_{in}$$

Supposing

(i) The feedback component is a capacitor C ,

i.e., $Z_f = \frac{1}{j\omega C}$

(ii) The input component is a resistor R , $Z_a = R$

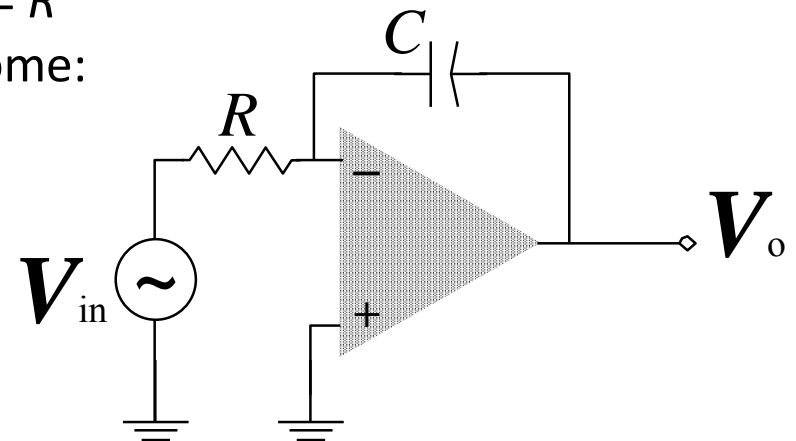
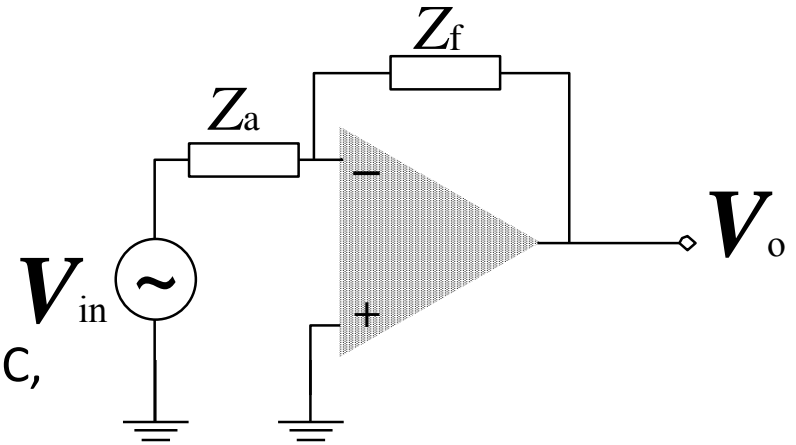
Therefore, the closed-loop gain (V_o/V_{in}) become:

$$v_o(t) = \frac{-1}{RC} \int v_i(t) dt$$

where $v_i(t) = V_i e^{j\omega t}$

What happens if $Z_a = 1/j\omega C$ whereas, $Z_f = R$?

Inverting differentiator

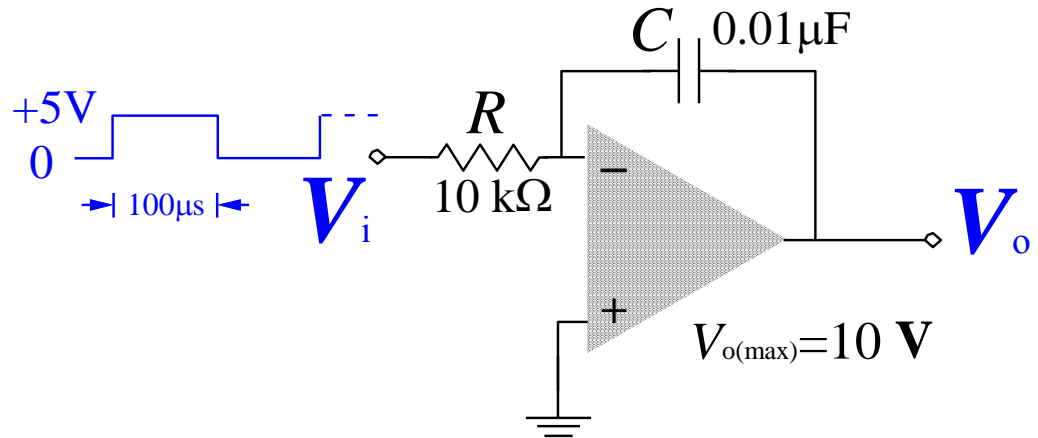


Op-Amp Integrator

Example:

- (a) Determine the rate of change of the output voltage.

- (b) Draw the output waveform.



Solution:

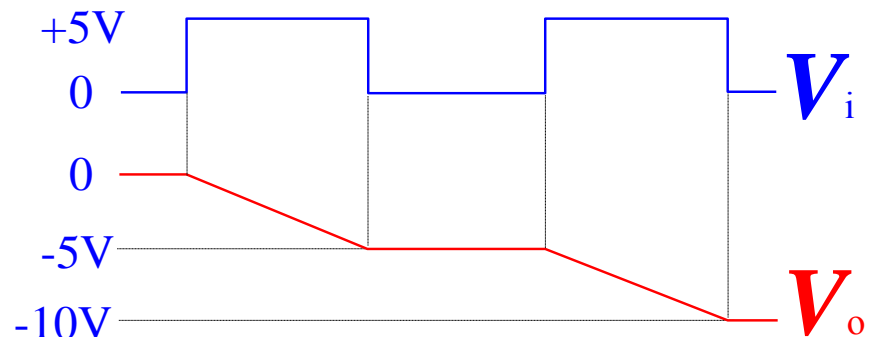
- (a) Rate of change of the output voltage

$$\frac{\Delta V_o}{\Delta t} = -\frac{V_i}{RC} = \frac{5 \text{ V}}{(10 \text{ k}\Omega)(0.01 \mu\text{F})}$$

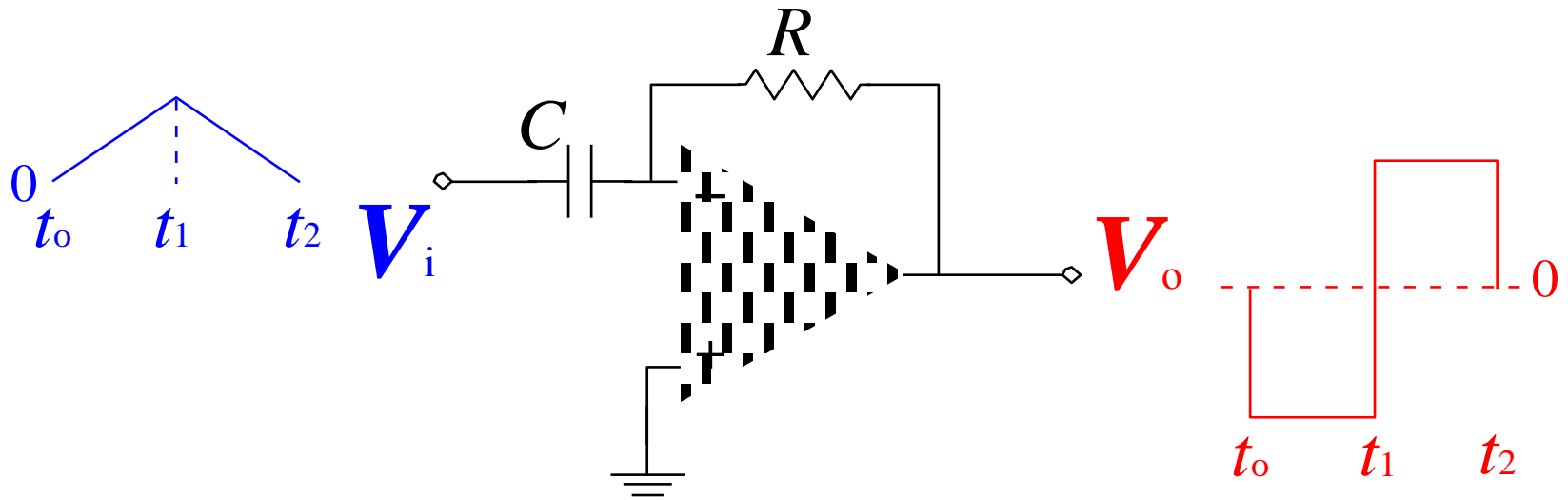
$$= -50 \text{ mV}/\mu\text{s}$$

- (b) In $100 \mu\text{s}$, the voltage decrease

$$\Delta V_o = (-50 \text{ mV}/\mu\text{s})(100 \mu\text{s}) = -5 \text{ V}$$



Op-Amp Differentiator



$$v_o = -\left(\frac{dV_i}{dt}\right)RC$$